Microelectronic Devices and Circuits- EECS105 Second Midterm Exam

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Costas J. Spanos
University of California at Berkeley
College of Engineering
Department of Electrical Engineering and Computer Sciences

Your Name:	Official Solutions		
1011111111111	(last)	(first)	
Your Signatur	e:		
l. Print and sign y	our name on this page befor	re you start.	
2. You are allowed	two, 8.5"x11" handwritter	sheet. No books or notes!	
3. A correct expres	sion is worth 70% of the cr	edit. The calculation gets yo	ou the rest.
1. You have until 4	1/25/03 to bring any grading	issues to Prof. Spanos' atte	ention.
	Problem 1	/48	
	Problem 2	/16	
	Problem 3	/16	
	Problem 4	/ 20	
	TOTAL 1=75	<u>σ:16</u> /100	

$$\begin{split} & \underline{MOS\ Device\ Data^1}\ (you\ may\ not\ have\ to\ use\ all\ of\ these...)}\\ & \mu_n C_{ox} = 50\mu A/V^2,\ \mu_p C_{ox} = 25\mu A/V^2,\ V_{Tn} = -V_{Tp} = 1V,\ Lmin = 2\mu m.\ V_{BS} = 0.\\ & \lambda_n = \lambda_p = 0.1V^{-1}\ when\ L = 1\mu m,\ and\ it\ is\ otherwise\ proportional\ to\ 1/L\\ & C_{ox} = 2.3fF/\mu m^2,\ C_{jn} = 0.1fF/\mu m^2,\ C_{jp} = 0.3fF/\mu m^2,\ C_{jswn} = 0.5fF/\mu m,\\ & C_{jswp} = 0.35fF/\mu m,\ C_{ovn} = 0.5fF/\mu m,\ C_{ovp} = 0.5fF/\mu m \end{split}$$

BJT Device Data¹ (you may not have to use all of these...) $β_r = β_o = 100, I_S = 10^{-17} A, \ V_{CE\ SAT} = 0.1 V, \ V_A = 25 V, \ \tau_F = 50 ps, \ C_{je} = 15 fF@V_{BE} = 0.7 V, \ C_{μ} = 10 fF@V_{BC} = -2.0 V$

Problem 1 of 4: Answer each question briefly and clearly. (6 pts each, 48 total)

1.1 What are the typical values of Rin, Rout for a decent CC voltage buffer?

				E .		
Rin	5-10 Ω	100-1000 Ω	10-20 kΩ	100-300 kΩ	$(1-10 \text{ M}\Omega)$	PA+Bo (2d/Toc/R)
	•			100 00010	1 10 110	11 2
Rout	$5-10 \Omega$	$(100-1000 \Omega$	$(10-20 \text{ k}\Omega)$	100-300 kΩ	1-10 M22	1/0 +
			7			" Par Bo
		~ ~	/			- 6

- 1.2 Why is it important to make the base of a BJT as short as possible? (choose one)
- So that the transistor occupies less space
- So that there are no ohmic losses in the base
- Y So that almost all injected minority carriers make it across
- So that holes and electrons stay uniformly distributed
- 1.3 What happens when the base of a BJT is very short? (choose one)
- □ The transonductance drops in value
- □ The beta drops in value
- The ro drops in value
- The Early voltage gets delayed

Provide a brief explanation of the mechanism responsible for the above:

Base width modulation due to the CB depletion region making the base were shorter (decreases (VA))

- 1.4 What happens when the channel of a MOSFET is very long? (choose one)
- The transconductance increases
- Velocity saturation kicks in
- ☐ The transistor gets too hot
- The ro increases in value

Provide a brief explanation of the mechanism responsible for the above:

(hannel learth modelarion is less of a problem for long transistors (decreases))

Except as indicated on the particular problem...

1.5 Why it is important to have a current supply with a huge (i.e. tens of MOhms) internal resistance when you design a current buffer utilizing a Common Base BJT? (Typical r_o for a BJT is 100kOhms)

Sur Delauxe a CB is meant to how a huye out put tesistance and you do not want the Zoc to limit it!

Too (It 9m (70/18)) 70]

1.6 Why it is NOT so important to have a current supply with an internal resistance greater than ~1MOhm when you are designing a Common Emitter amplifier? (Typical r₀ for a BJT is 100kOhms)

Recause roc your in parallel to ro
and ro limits the Rous value. A very
large roc will be masted...

Rous: rollroc

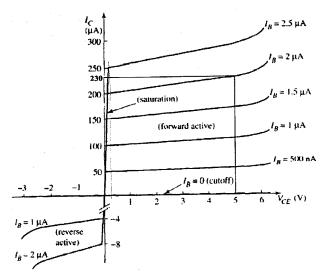
1.7 Draw the small signal model of the forward biased diode. Name all distinct capacitive components.

1.8 Provide a brief definition of the transition frequency ω_t of the BJT transistor.

the frequency or which a CE amp with shorted output has a current gain of -1.

Problem 2 of 4: Answer each question briefly and clearly. (16 points)

From the graph below, please estimate β_F and β_R , as well as r_o . Also, estimate g_m r_π and $V_{\text{CESAT min}}$ when $I_B = 2\mu A$ at room temperature ($V_{th} = 0.025 mV$. Do NOT use the default BJT values on the front page of this exam for this question. Hint: in reverse active mode the role of the collector and emitter are reversed. Mark the graph to clarify your method as needed.)

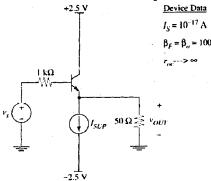


expression	value & units	
BF = Ic/IB = 200 pA/2pA	100	7
$\beta_R = I_{\epsilon/I_{1\beta}} = I_{\epsilon-I_{\beta}/I_{\beta}} = \frac{g-2}{2}$	3	٦.
ro = \$\frac{1}{5000} = \frac{1}{3000} = \frac{5V}{3000} = 5V	167 KO	3

Assuming $I_B = 2\mu A$:

g _m =	Te Br. Ju	200µA =	25.10-2 = 08.10	8m5	3
$r_{\pi} =$	BF/9m=	100./8.10	3	12.5 Kg	3
V _{CE SA}	Tmin = ~0.2V	Cfrom the y	rae h)	0.2V	3

Problem 3 of 4: Answer each question briefly and clearly. (16 points)



You are given the above Common Collector amplifier, with an ideal current supply source, and with values as shown above. Please do the following:

3.1 Calculate the output voltage swing, assuming that the I_{SUP} needs a minimum of 0.5Volts across it in order to operate properly, and V_{CE SATmin} is 0.1V.

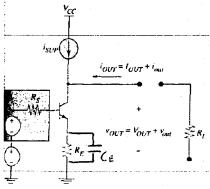
Voutmax = limited by BST staying in FAM
Voutman : limited by Isup staying functional

Expression Value $v_{OUTmax} = V^{+} - V_{CESA7.nein} = 2.5 - 0.4$ $v_{OUTmin} = V^{-} + V_{Sup_{unin}} = -2.5 + 0.5$

3.2 Draw the equivalent small signal circuit of the common collector amplifier shown on the preceding page.

Problem 4 of 4: Answer each question briefly and clearly. (20 points)

You are given a common emitter amplifier with R_{E} (degeneration emitter resistor), and an extra capacitor C_{E} connected across that resistor.



Recall that without the capacitance C_E , the formulae for this amplifier are:

 G_{m}

(1 ± 0 RJ

 $r_{\mathbf{g}}[1+g_{\mathbf{pq}}R_{\mathbf{E}}] \parallel r_{\mathbf{ec}}$

4.1 Write expressions for R_{in} , R_{out} and G_m as a function of ω when C_E is present (neglect all the transistor small signal capacitances).

expression

$$G_{m} = \frac{9m}{|+9mZ_{E}|} = \frac{9m(1+jwfeCE)}{|+jwfeCE+9mFE|}$$

$$R_{in} = 2n(1+9mZ_{E}) = 2n(1+\frac{9mFE}{|+jwFeCE|})$$

$$R_{out} = 2n(1+9mZ_{E}) + 2n(1+\frac{9mFE}{|+jwFeCE|}) + 2n(1+\frac{9mFE}{|$$

4.2 Given the following values: $R_E = 10 kOhms$, $g_m = 0.9 mS$, and $C_E = 100 nF$, write the expression $G_m(\omega)$, restate it to show its pole(s) and zero(es) and plot it on a Bode Plot, whose vertical axis is in units of $20log_{10}(|Gm|/mS)$. (These units are actually db-mS with reference of 1mS, i.e. a value of 1mS will show as 0db, 10mS will show as 20db, etc.)

Gm =
$$\frac{9m}{1+9m(RE|I]} = \frac{9m(I+jwRE(E))}{1+jwRE(E+9mRE)} = \frac{9m}{1+9mRE} = \frac{1+jwRE(E)}{1+jwRE(E)}$$
 $Z_1 = \frac{1}{RE(E)}$
 $P_1 = \frac{1+9mRE}{RE(E)}$

Expression $G_{m}(\omega) = \frac{g_{m}}{14g_{m}} \left[\frac{1+\int w \, \text{Rele}}{1+\int w \, \text{Rele}} \right]$ $G_{m} \text{ when } \omega \text{ very low} = \frac{g_{m}}{1+g_{m}} \left[\frac{g_{m}}{1+g_{m}} \right]$ $G_{m} \text{ when } \omega \text{ very high} = g_{m}$ $0.9 \text{ m/s} \quad \sqrt{-210 \text{ bm/s}}$

