

Final
EE40
Spring 2015

NAME: _____ SSID: _____

Instructions

Read all of the instructions and all of the questions before beginning the exam.

There are 6 problems in this exam. The total score is 150 points. Points are given next to each problem to help you allocate time. Do not spend all your time on one problem.

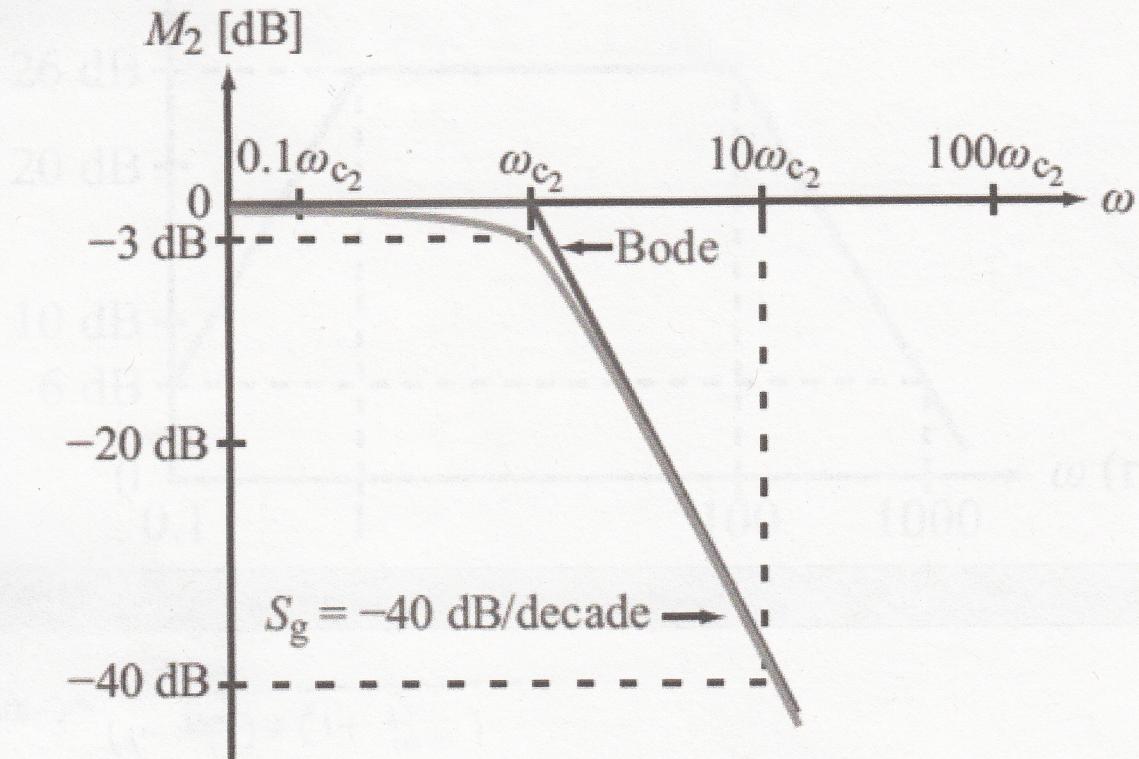
Unless otherwise noted on a particular problem, you must show your work in the space provided, on the back of the exam pages or in the extra pages provided at the back of the exam.

Be sure to provide units where necessary.

GOOD LUCK!

PROBLEM	POINTS	MAX
1		?
2		?
3		?
4		?
5		?
6		?

Problem 1 Warm up (*n* points each, *m* points total)
 Consider the following magnitude Bode plot:



a) Provide a transfer function that can generate this magnitude Bode plot:

Solution:

$$H(\omega) = \frac{1}{(1 + \frac{\omega}{\omega_{c2}})^2}$$

two poles at ω_{c2}

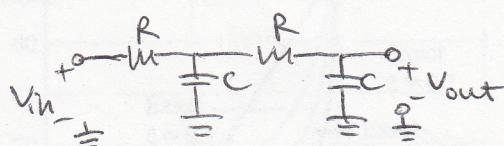
$$H = \frac{1}{(1 + \frac{\omega}{\omega_{c2}})^2}$$

$$\log |H(\omega=0)| \approx \log(1) = 0$$

clearly label V_{in} and V_{out}

b) Provide a circuit which has the Bode magnitude plot above. There's no need to provide numerical values for the components

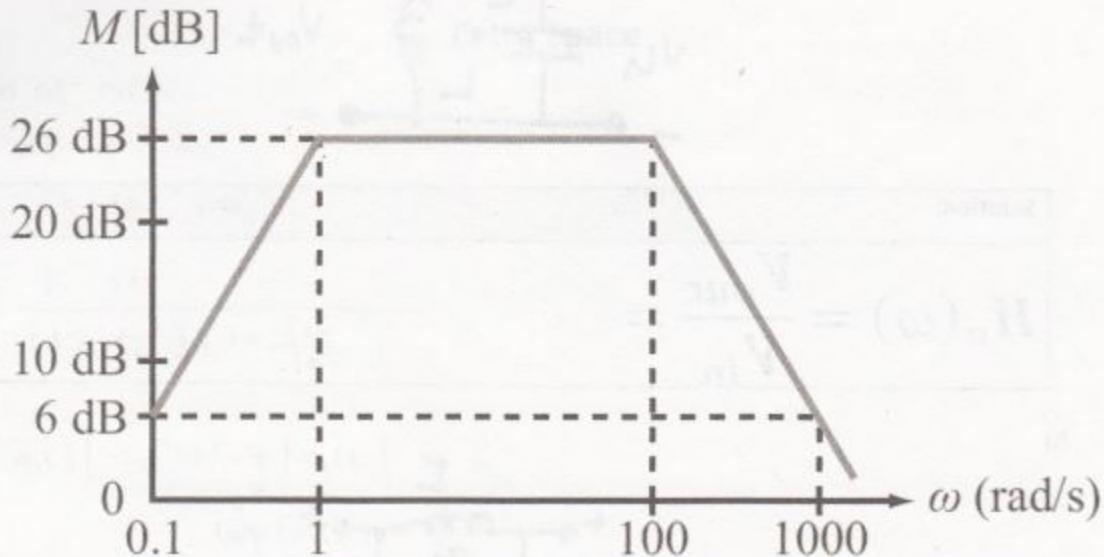
Solution:



$$\frac{V_{out}(\omega)}{V_{in}(\omega)} = H(\omega)$$

hint: $20 \log 2 = 6$.

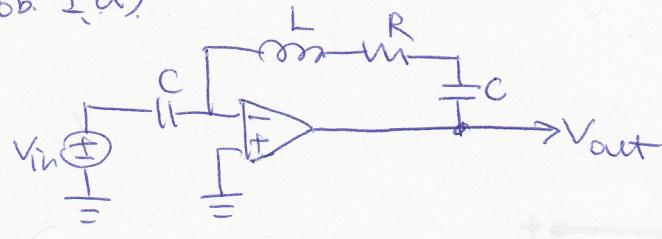
c) Provide a transfer function that produces the Bode plot below:



Solution:

$$H(\omega) = \frac{j2\omega}{(1 + j\omega)} \cdot (1 + j\frac{\omega}{100})$$

Prob. 1.d)



$$-\frac{V_{out}}{V_{in}} = s^2 LC + sRC + 1.$$

Extra Space

c). zero at origin.

one pole at 1 rad/s

one pole at 100 rad/s

$$H(\omega) = \frac{K \cdot j\omega}{(1 + \frac{j\omega}{1}) (1 + \frac{j\omega}{100})}$$

$$20 \log |H(\omega=0.1)| = 20 \log |0.1K| = 6$$

$$\log |0.1K| = \frac{6}{20}$$

$$0.1K = 2$$

$$K = 20.$$

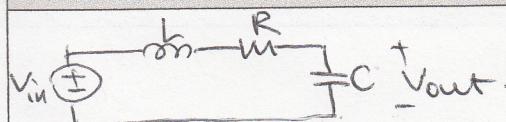
Problem 2 (n points)

We want to design the *simplest* voltage filter with:

- a center frequency, $\omega_{center} = 1 \text{ GHz rad/s}$.
- a voltage transfer function magnitude at the center frequency, $|\hat{A}_v(\omega_{center})| = 100 \text{ dB}$
- which uses only linear passive components

a) Please draw the circuit below inside the box: (clearly label V_{in} and V_{out})

Solution:



$$\frac{V_{out}}{V_{in}} = \frac{\frac{1}{sC}}{sL + R + \frac{1}{sC}} = \frac{1}{s^2LC + sRC + 1}$$

$$\text{at } \omega_0 = \frac{1}{\sqrt{LC}} = 10^9$$

$$20 \log(10^5) = 100 \text{ dB}$$

$$\left| \frac{V_{out}}{V_{in}} (\omega_0) \right| = \frac{1}{\frac{RC}{\sqrt{LC}}} = \frac{\sqrt{L}}{R\sqrt{C}} = 10^5$$

b) Please list any relevant component values below:

Solution:

any R, L, C combination that fulfills the following:

$$\begin{cases} \frac{1}{\sqrt{LC}} = 10^9 \\ \frac{\sqrt{L}}{R\sqrt{C}} = 10^5 \end{cases}$$

Works.

$$\text{e.g. } \frac{1}{\sqrt{LC}} = 10^9 \Rightarrow C = 1 \text{ pF} = 10^{-12} \quad L = 10^{-6} \text{ H} = 1 \text{ mH}$$

$$\frac{1}{R} \sqrt{\frac{10^{-6}}{10^{-12}}} = \frac{1}{R} 10^3 = 10^5 \Rightarrow R = 10 \text{ m}\Omega$$

are sinusoidal.

c) Is the circuit resonant? If so, what is the Q? If not, why not?

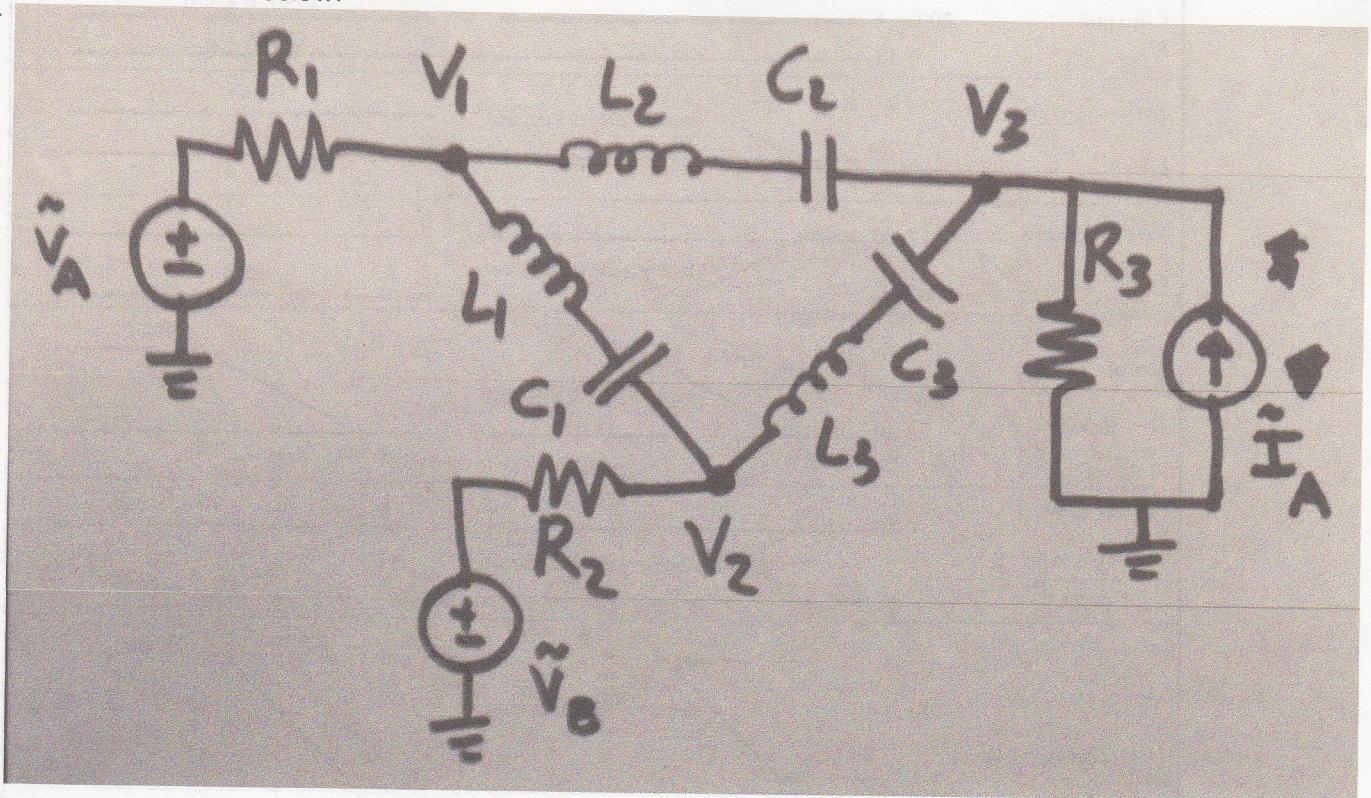
Solution:

It resonates at ω_{center}

$$Q = \frac{\omega_0 L}{R} = \frac{10^9 \cdot 10^{-6}}{10^{-2}} = 10^5$$

Problem 3 Nodal (*n* points)

Consider the circuit below.



$$R_1 = 100 \Omega \quad L_1 = 1 \mu\text{H} \quad C_1 = 1 \mu\text{F}$$

$$R_2 = 100 \Omega \quad L_2 = 5 \mu\text{H} \quad C_2 = 200 \text{nF}$$

$$R_3 = 100 \Omega \quad L_3 = 10 \mu\text{H} \quad C_3 = 100 \text{nF}$$

The three sources are provided as phasors:

$$\tilde{V}_A = 1 \text{ V}, 0^\circ \text{ at } \omega = 10^6 \text{ rad/s}$$

$$\tilde{V}_{A_B} = 1 \text{ V}, 0^\circ \text{ at } \omega = 10^6 \text{ rad/s}$$

$$\tilde{I}_A = 1 \text{ A}, 0^\circ \text{ at } \omega = 10^6 \text{ rad/s}$$

Solution:

$$V_1 = V_2 = V_3(t) = 34 \cos(10^6 t)$$

$$\frac{V_A - V_1}{R_1} = \frac{V_1 - V_3}{sL_2 + \frac{1}{sC_2}} + \frac{V_1 - V_2}{sL_1 + \frac{1}{sC_1}}$$

$$\frac{V_1 - V_3}{sL_2 + \frac{1}{sC_2}} + \frac{V_2 - V_3}{sL_3 + \frac{1}{sC_3}} + I_A = \frac{V_3}{R_3}$$

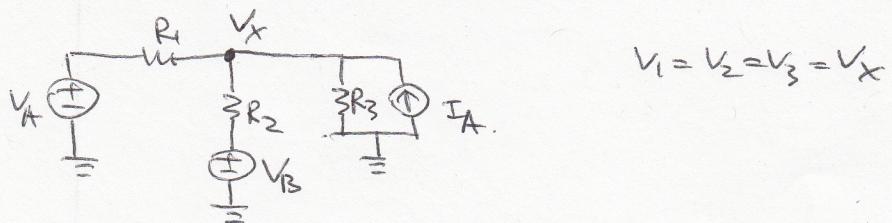
$$\frac{V_1 - V_2}{sL_1 + \frac{1}{sC_1}} + \frac{V_B - V_2}{R_2} = \frac{V_2 - V_3}{sL_3 + \frac{1}{sC_3}}$$

$$sL_1 + \frac{1}{sC_1} = j10^6 \cdot 10^{-6} + \frac{1}{j10^6 \cdot 10^{-6}} = j + \frac{j}{f} = 0.$$

$$sL_2 + \frac{1}{sC_2} = j10^6 \cdot 5 \cdot 10^{-6} + \frac{1}{j10^6 \cdot 200 \cdot 10^{-9}} \\ = 5j + \frac{5000}{j200} = 0.$$

$$sL_3 + \frac{1}{sC_3} = j10^6 \cdot 10 \cdot 10^{-6} + \frac{1}{j10^6 \cdot 100 \cdot 10^{-9}} = j10 + \frac{1000}{j10} = 0$$

equivalent ckt:



$$\frac{V_A - V_x}{R_1} + \frac{V_B - V_x}{R_2} + I_A = \frac{V_x}{R_3}$$

$$\frac{\frac{V_A}{R_1} + \frac{V_B}{R_2} + I_A}{\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}} = V_x$$

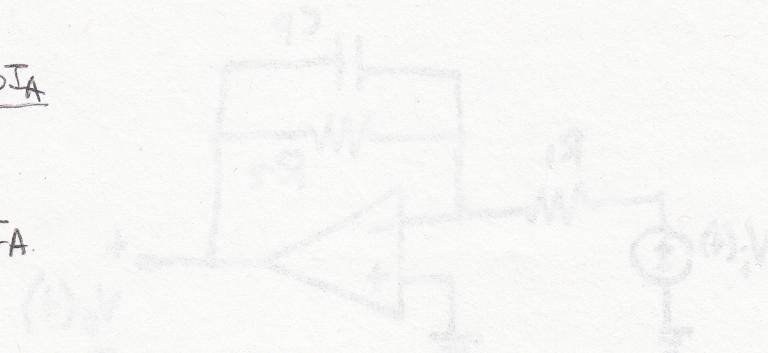
$$V_x = \frac{\left(\frac{1}{R_1} + \frac{1}{R_2}\right)V_A + I_A}{\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}}$$

$$V_x = \frac{\frac{2}{100} V_A + I_A}{\frac{3}{100}}$$

$$\text{II} \quad \frac{2V_A + 100I_A}{3}$$

$$\text{II} \quad \frac{2}{3}V_A + \frac{100}{3}I_A$$

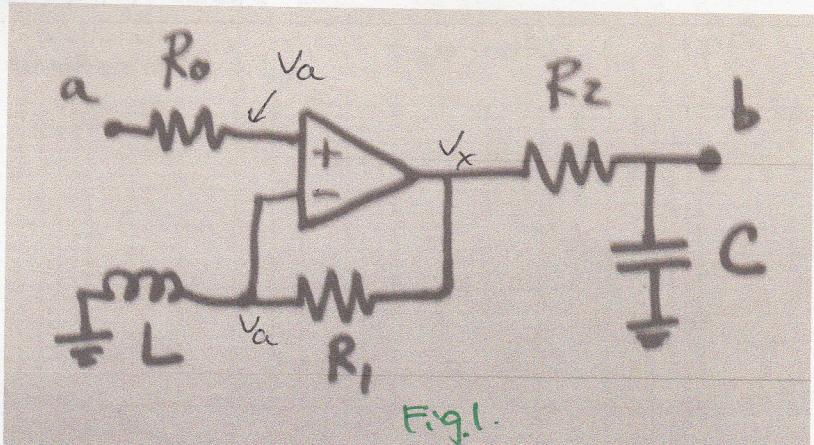
$$\begin{aligned} \text{III} \\ V_x(t) &= \frac{2}{3} \cos(\omega_0 t) + \frac{100}{3} \cos(\omega_0 t) \\ &= 34 \cos(10^6 t) \end{aligned}$$



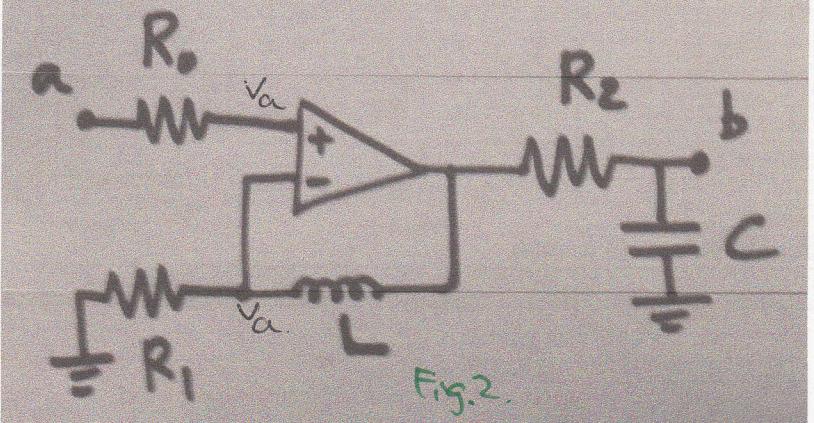
Problem 4 (n points)

Consider the two circuits below.

$$\left\{ \begin{array}{l} \frac{V_a}{S_L} = \frac{V_x - V_a}{R_1} \\ \frac{V_x - V_b}{R_2} = V_{b S C} \end{array} \right.$$



$$\left\{ \begin{array}{l} \frac{V_a}{R_1} = \frac{V_x - V_a}{S_L} \\ \frac{V_x - V_b}{R_2} = V_{b S C} \end{array} \right.$$



a) Provide SYMBOLIC transfer functions for BOTH circuits. (a as input port and b as output port).

Transfer function for the left circuit: (Fig. 1)	Transfer function for the right circuit: (Fig. 2)
$\frac{V_a}{S_L} \neq \frac{j\omega R_1 (1 + j\omega R_2 C)}{1 + j\omega L/R_1}$ $\frac{V_a}{S_L} \neq \frac{j\omega R_1 (1 + j\omega R_2 C)}{1 + j\omega L/R_1}$ $\frac{V_b}{V_a} = \frac{1 + S_L / R_1}{S L R_1 (1 + S R_2 C)}$ $\frac{V_b}{V_a} = \frac{10^3 (1 + \frac{S}{10^3})}{S (1 + \frac{S}{10^3})}$ $\frac{V_a}{V_b} = \frac{j\omega \times 10^{-9} (1 + j\omega \times 50R_2)}{1 + j\omega / 10^3}$	$\frac{V_b}{V_a} = \frac{1 + \frac{S_L}{R_1}}{1 + S R_2 C}$
	$\frac{V_b}{V_a} = \frac{1 + \frac{j\omega}{10^3}}{1 + \frac{j\omega}{10^3}}$

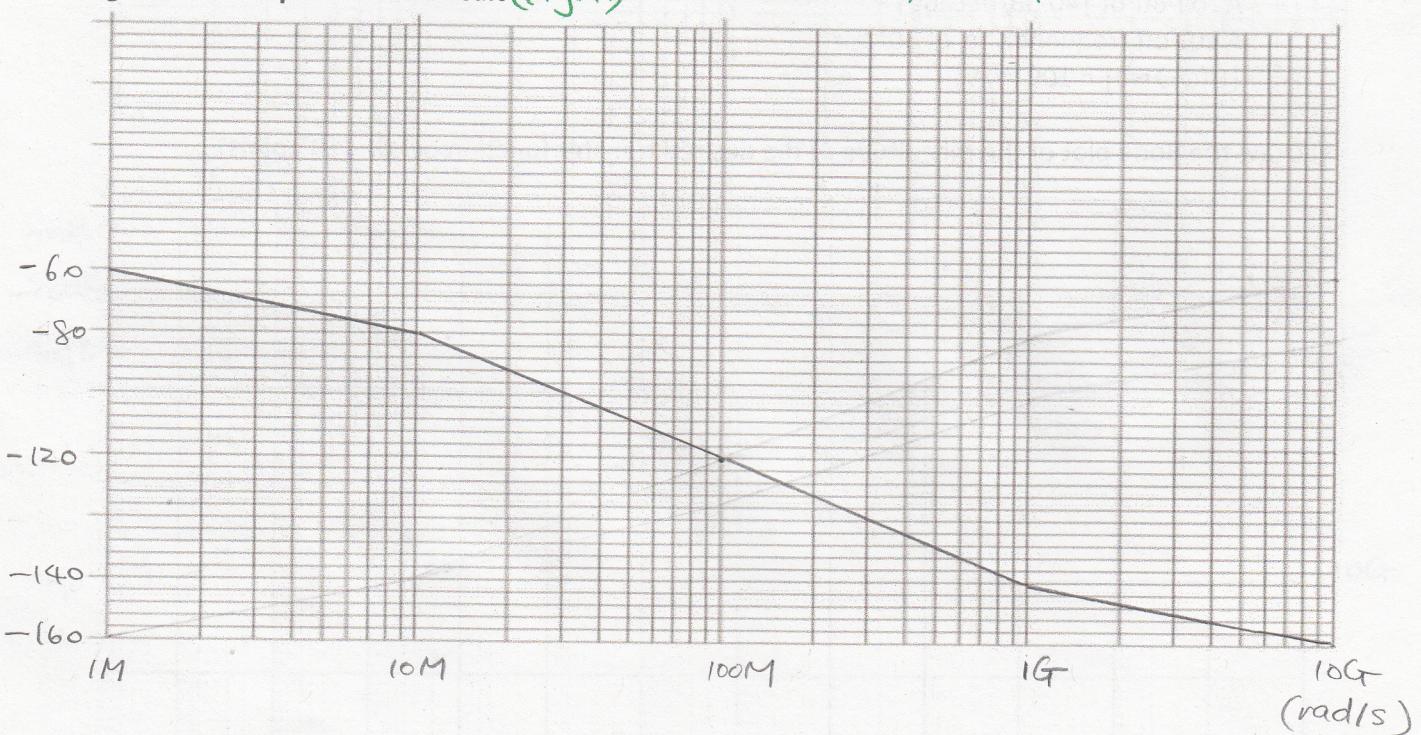
$$20 \log |H(j\omega=1M)| = 20 \log 10^{-3} = -60 \text{dB}$$

b) Provide Bode magnitude and phase plots for both circuits: (n points)

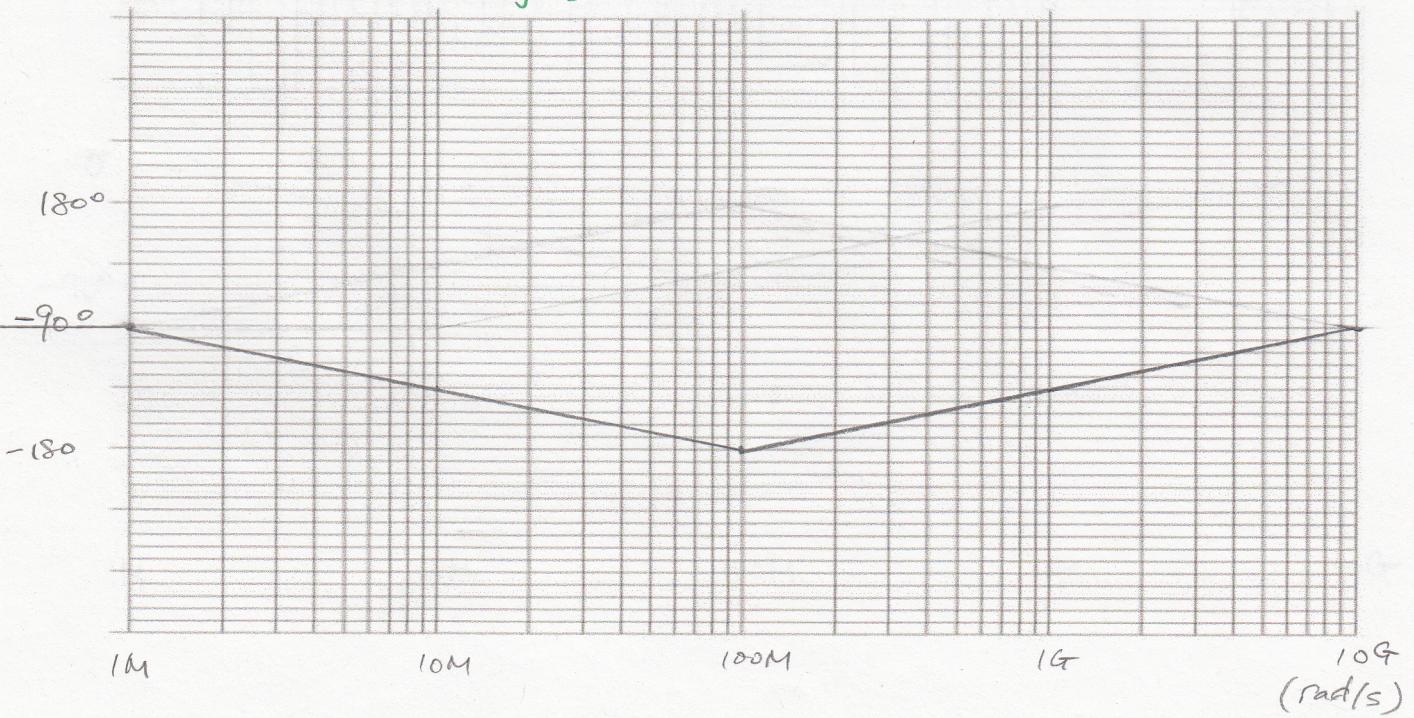
$$R_0 = 100 \Omega, R_1 = 1 \text{ k}\Omega, L = 1 \mu\text{H}, R_2 = 1 \text{ M}\Omega, C = 20 \text{ pF}$$

Magnitude Bode plot for left circuit (Fig. 1.)

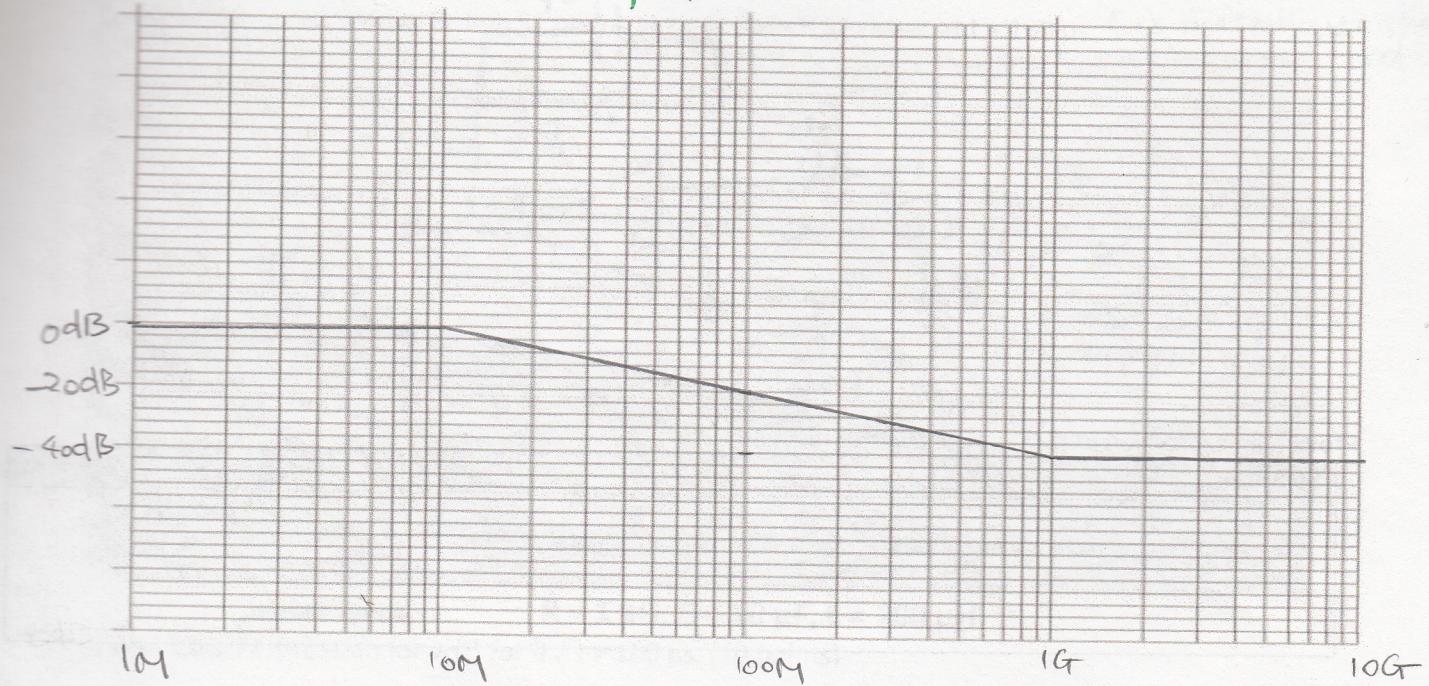
$$R_2 = 100 \text{ k}\Omega, C = 1 \text{ pF}$$



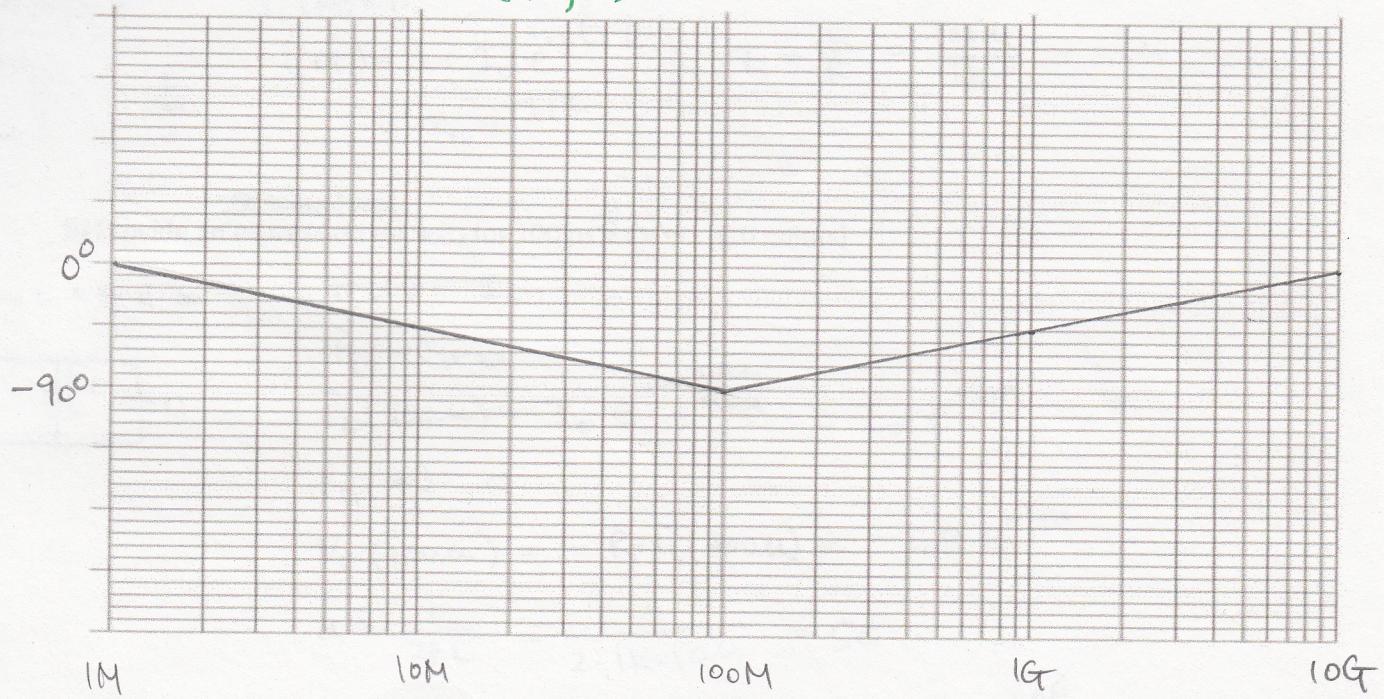
Phase Bode plot for left circuit (Fig. 1.)



Magnitude Bode plot for right circuit (Fig.2.)

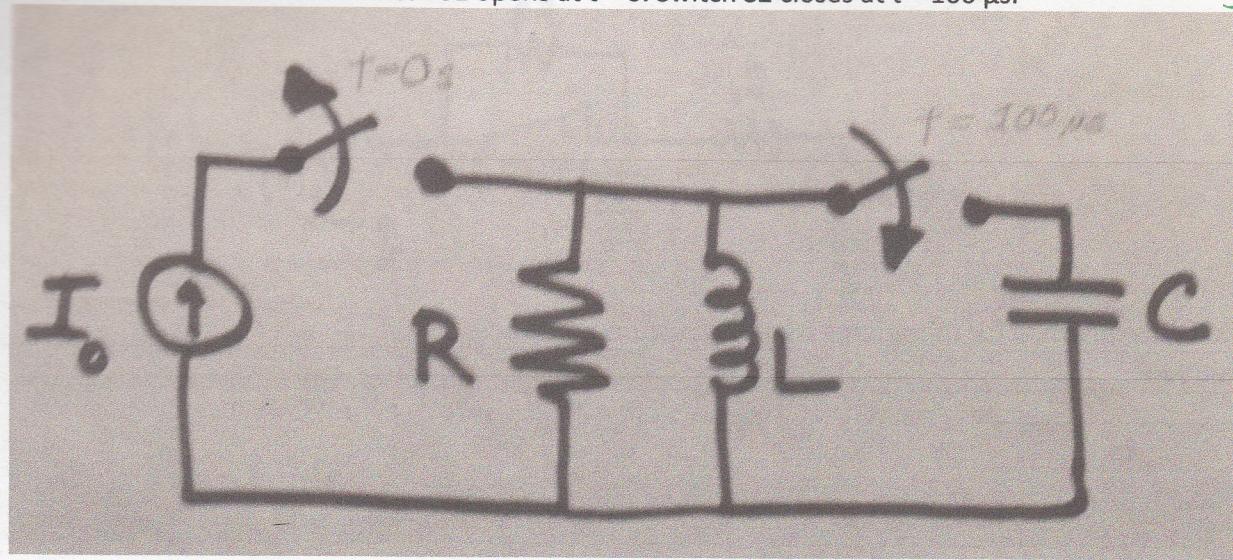


Phase Bode plot for right circuit (Fig.2)



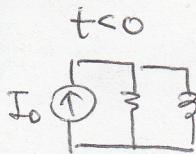
Problem 5 (25 points)

Consider the circuit below. Switch S1 opens at $t = 0$. Switch S2 closes at $t = 100 \mu\text{s}$. C is initially uncharged.

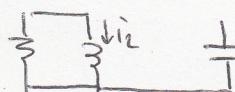


$$R = 1 \text{ k}\Omega, C = 10 \mu\text{F}, L = 100 \mu\text{H}$$

a) Provide an expression for $v_C(t)$ for $0 > t > 100 \mu\text{s}$. (10 points)



$0 < t < 100 \mu\text{s}$



$$V_C(t) = 0 \text{ V}$$

$$i_L(0) = I_0$$

$$i_L(\infty) = 0$$

$$i_L(t) = I_0 e^{-t/\tau}, \quad \tau = \frac{L}{R} = \frac{100\mu\text{H}}{1\text{k}} = 0.1\mu\text{s}$$

$$= I_0 e^{-t/\tau}$$

b) Provide an expression for $v_C(t)$ for $100 \mu\text{s} < t > \infty$ (10 points)

$100 \mu\text{s} < t < \infty$

$$i_L(0) = I_0$$

$$i_L(\infty) = 0$$

$$i_L(100\mu\text{s}) = I_0 e^{-\frac{100\mu\text{s}}{0.1\mu\text{s}}} = I_0 e^{-100}$$

$$i_L(\infty) = 0$$

$$V_L(100\mu\text{s}) = -R \cdot i_L(100\mu\text{s}) = -I_0 R e^{-100}$$

$$\alpha = \frac{1}{2RC} = \frac{1}{2 \cdot 1\text{k} \cdot 10\mu\text{s}} = 50$$

$$\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{100\mu\text{s} \cdot 10\mu\text{s}}} = \frac{10^6}{\sqrt{1000}}$$

$$50^2 < \frac{(10^6)^2}{1000}$$

$\therefore \alpha < \omega_0 \Rightarrow \text{underdamped}$

Extra Space

$$w_d = \sqrt{w_0^2 - \alpha^2} \quad \text{Known}$$

$$D_1 = i_L(100u) - i_L(\infty) = I_0 e^{-1m} \quad \text{Known}$$

$$D_2 = \frac{\frac{1}{L} V_L(0) + \alpha [i_L(0) - i_L(\infty)]}{w_d} \quad \text{Known.}$$

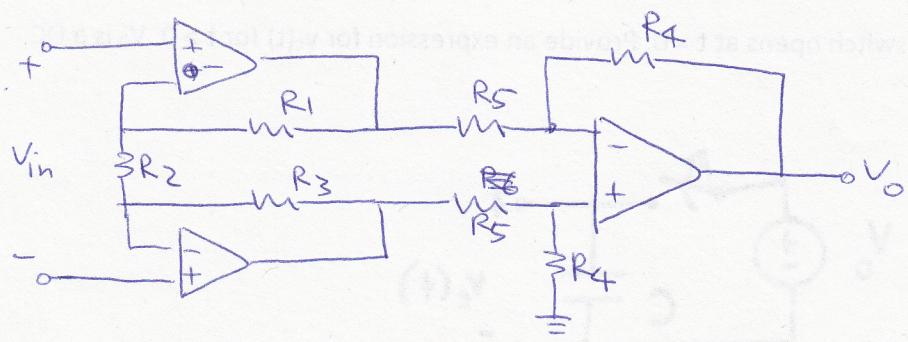
$$\therefore i_L(t) = e^{-\alpha t} (D_1 \cos w_d t + D_2 \sin w_d t)$$

$$V_L(t) = \frac{di_L}{dt} = L(-\alpha) e^{-\alpha t} (D_1 \cos w_d t + D_2 \sin w_d t)$$

$$i_L(t) = e^{-\alpha(t-100u)} (D_1 \cos w_d(t-100u) + D_2 \sin w_d(t-100u))$$

$$\begin{aligned} V_L(t) &= L \frac{di_L}{dt} = L \cdot e^{100u} \cdot (-\alpha) e^{-\alpha(t)} (D_1 \cos w_d(t-100u) + D_2 \sin w_d(t-100u)) \\ &\quad + L e^{100u} \cdot e^{-\alpha t} \cdot (-D_1 w_d \sin w_d(t-100u) + D_2 w_d \\ &\quad \cdot \cos w_d(t-100u)) \end{aligned}$$

Prob 6. a)



$$\frac{V_o}{V_{in}} = \left(\frac{R_4}{R_5} \right) \left(\frac{R_1 + R_2 + R_3}{R_2} \right) = 1000$$

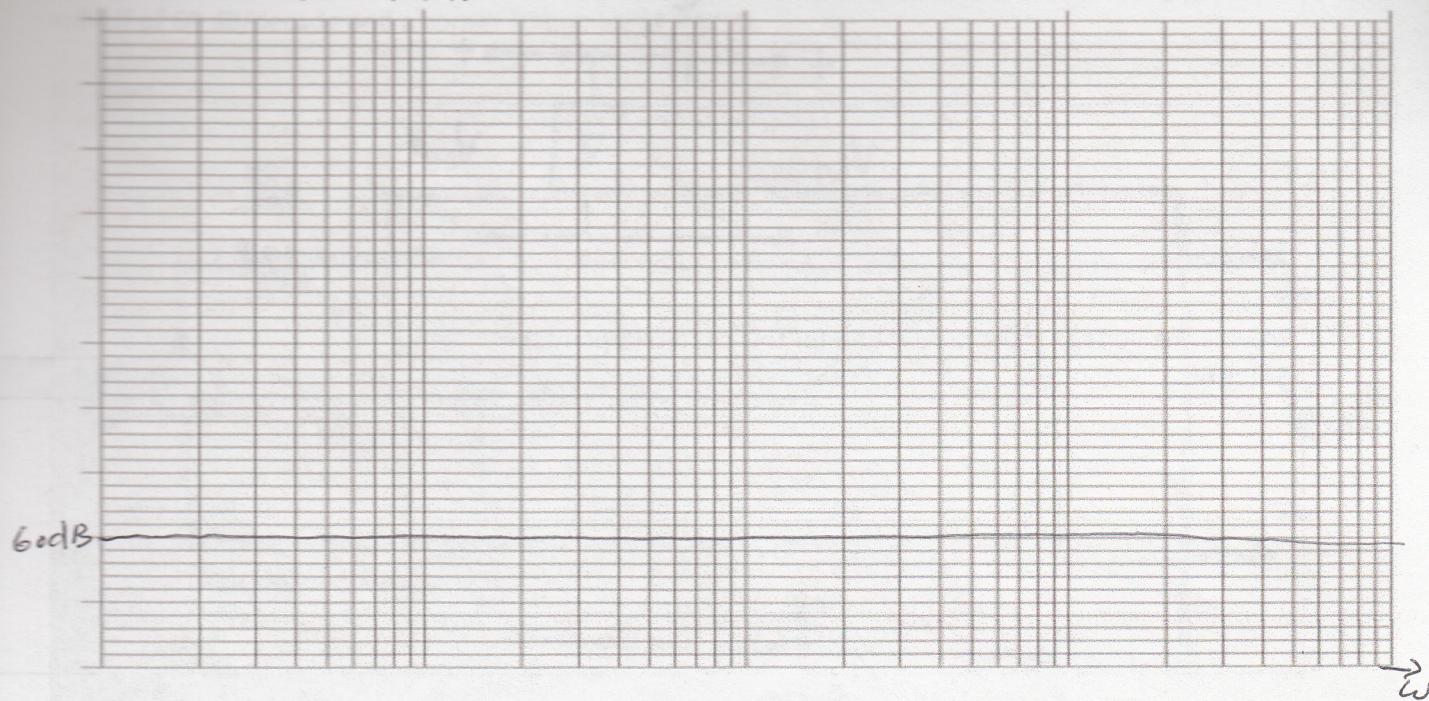
e.g. $R_4 = \frac{100}{3} k\Omega$, $R_5 = 100\Omega$

$$R_1 = R_2 = R_3 = 100\Omega$$

of the

✓ magnitude

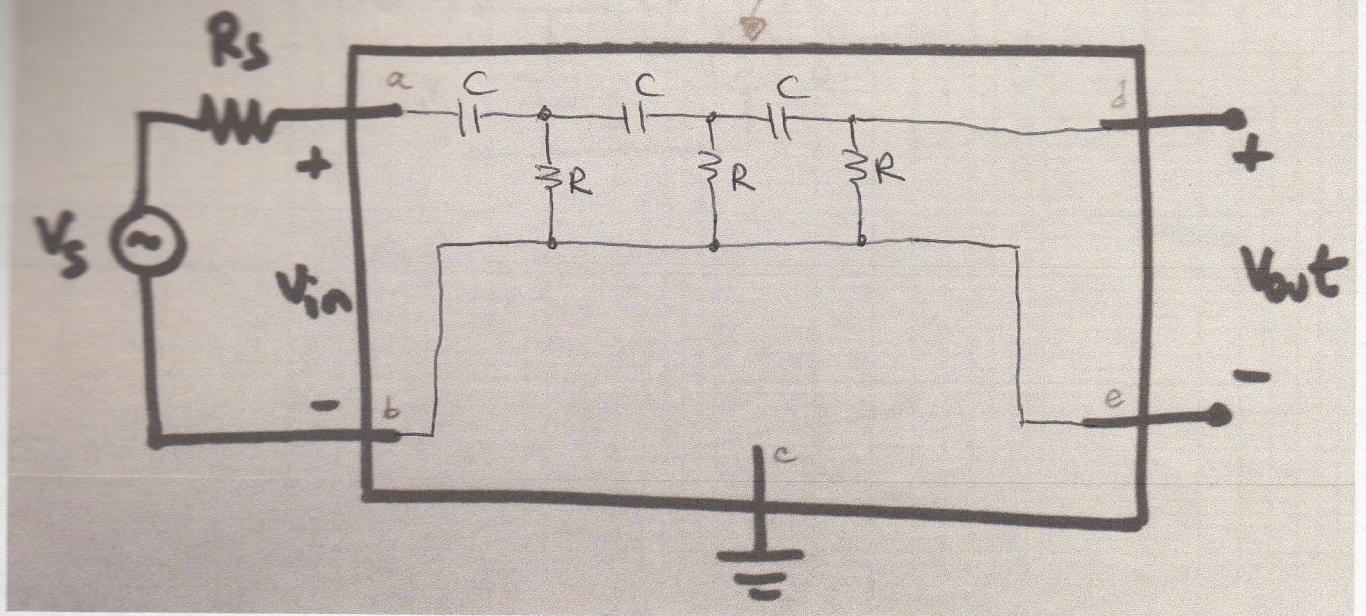
(ii) Provide the Bode plot of $|H(\omega)|$.



$$20 \log |H(\omega)| = 20 \log 10^3 = 60 \text{ dB}$$

Q) Instead of amplifying, we want to filter out 60 Hz noise. Design a passive non-resonant highpass filter with a roll-off of 60 dB/decade and a corner frequency of 300 Hz.

Draw your circuit in the box



$$\left\{ \frac{H(\omega)}{R + \frac{1}{sC}} \right\}^3 = \left\{ \frac{sRC}{1+sRC} \right\}^3$$

$$RC = \frac{1}{300 \times 2\pi}$$

$$20 \log |H(j\omega)|$$

$$20 \log |H(j\omega)| = 60 \log RC = 60 \log \frac{1}{300 \times 2\pi}$$

$$20 \log |H(j\omega)| = 60 \log \frac{1}{300}$$

Magnitude

a) Provide the Bode plot of $|\hat{H}(\omega)|$.

